

Chapter 4

Linear Estimation Theory

- Virtually all branches of science, engineering, and social science for data analysis, system control subject to random disturbances or for decision making based on incomplete information call for estimation theory.
- Many estimation problem can be formulated as a minimum norm problem in Hilbert space.
- The projection theorem can be applied directly to the area of statistical estimation.
- There are a number of different ways to formulate a statistical estimation.
 - ◊ Least squares.
 - ◊ Maximum likelihood
 - ◊ Bayesian techniques.
- When all variables are Gaussian statistics, these techniques produce linear equations.

Preliminaries

- If x is a real-valued random variable,

- ◇ The *probability distribution* P of the variable x is defined to be

$$P(\xi) = \text{Prob}(x \leq \xi).$$

- ◇ The “derivative” $p(\xi)$ of the probability distribution $P(\xi)$ is called the *probability density function (pdf)* of the variable x , i.e.,

$$P(\xi) = \int_{-\infty}^{\xi} p(x)dx.$$

- ▷ Note that

$$\int_{-\infty}^{\infty} p(x)dx = 1.$$

- ▷ $p(\xi) \geq 0$ for all ξ .

- The expected value of any function g of x is defined to be

$$\mathcal{E}[g(x)] := \int_{-\infty}^{\infty} g(\xi)p(\xi)d\xi.$$

- ◇ The *expected value* of x is $\mathcal{E}[x]$.

- ◇ The *variance* of x is $\mathcal{E}[(x - \mathcal{E}[x])^2]$.

- For random vector $\mathbf{x} = [x_1, \dots, x_n]^\top$,

- ◇ There is a *joint probability distribution* P defined by

$$P(\xi_1, \dots, \xi_n) = \text{Prob}(x_1 \leq \xi_1, \dots, x_n \leq \xi_n).$$

- ◇ The *covariance matrix* $\text{cov}(\mathbf{x})$ is defined by

$$\text{cov}(\mathbf{x}) = \mathcal{E} [(\mathbf{x} - \mathcal{E}[\mathbf{x}])(\mathbf{x} - \mathcal{E}[\mathbf{x}])^\top].$$

- ◇ Two random variables x_i and x_j are said to be uncorrelated or stochastically independent if

$$\mathcal{E}[(x_1 - \mathcal{E}[x_1])(x_2 - \mathcal{E}[x_2])] = \mathcal{E}[x_1 - \mathcal{E}[x_1]]\mathcal{E}[x_2 - \mathcal{E}[x_2]].$$

Least Squares Model

- This is a familiar subject as we have seen in many occasions.
- This problem is not a statistical one.
- It amounts to approximating a vector $\mathbf{y} \in \mathbb{R}^m$ by a vector lying in the column space of $W \in \mathbf{R}^{m \times n}$ and $n < m$.

◊ We assume a linear model that the response \mathbf{y} is related to the input $\boldsymbol{\beta}$ linearly, i.e.,

$$\mathbf{y} = W\boldsymbol{\beta}.$$

- ◊ We would like to recover $\boldsymbol{\beta}$ from observed \mathbf{y} . (Would it be a linear relationship?)
 - ◊ We are not assuming that the observed \mathbf{y} carries errors.
- It would be interesting to compare the least squares setting with those with random noises.

Least Squares Formulation

- Given

- ◊ A known matrix $W \in \mathbb{R}^{m \times n}$, $n < m$.

- ◊ An observation vector $\mathbf{y} \in \mathbb{R}^m$.

Find $\hat{\boldsymbol{\beta}} \in \mathbb{R}^n$ such that $\|\mathbf{y} - W\hat{\boldsymbol{\beta}}\|$ is minimized over all $\boldsymbol{\beta} \in \mathbb{R}^n$.

- By the projection theorem, the solution exists and is unique.

- The normal equation is given by

$$W^{\top}(\mathbf{y} - W\hat{\boldsymbol{\beta}}) = 0.$$

- If W has linear independent columns, then

$$\hat{\boldsymbol{\beta}} = \underbrace{(W^{\top}W)^{-1}W^{\top}}_K \mathbf{y}.$$

- ◊ Note that the optimal solution $\hat{\boldsymbol{\beta}}$ is related to \mathbf{y} linearly.

Gauss-Markov Model

- A more realistic model in an experiment is

$$\mathbf{y} = W\boldsymbol{\beta} + \boldsymbol{\epsilon}.$$

- ◇ $W \in \mathbb{R}^{m \times n}$ is known.
 - ◇ $\boldsymbol{\epsilon} \in \mathbb{R}^m$ is a random vector with zero mean and covariance $\mathcal{E}(\boldsymbol{\epsilon}\boldsymbol{\epsilon}^\top) = Q$.
 - ◇ \mathbf{y} represents the outcome of inexact measurements in \mathbb{R}^m .
- Want to estimate unknown parameter vector $\boldsymbol{\beta} \in \mathbb{R}^n$ from $\mathbf{y} \in \mathbb{R}^m$ using

$$\hat{\boldsymbol{\beta}} := K\mathbf{y}$$

with K an unknown matrix in $\mathbb{R}^{n \times m}$.

- Suppose the approximation is measured by minimizing the expected value of the error, i.e.,

$$\min_{K \in \mathbb{R}^{n \times m}} \mathcal{E}[\|\hat{\boldsymbol{\beta}} - \boldsymbol{\beta}\|^2]$$

- ◇ Since \mathbf{y} carries random noise, it is a random vector.
 - ◇ Both estimate $\hat{\boldsymbol{\beta}}$ and the difference $\hat{\boldsymbol{\beta}} - \boldsymbol{\beta}$ are random vectors.
 - ◇ The statistics of these random vectors are determined by those of $\boldsymbol{\epsilon}$ and K .

Gauss-Markov Estimate

- Observe that

$$\begin{aligned}\mathcal{E}[\|\hat{\boldsymbol{\beta}} - \boldsymbol{\beta}\|^2] &= \mathcal{E}[\langle K(W\boldsymbol{\beta} + \boldsymbol{\epsilon}) - \boldsymbol{\beta}, K(W\boldsymbol{\beta} + \boldsymbol{\epsilon}) - \boldsymbol{\beta} \rangle] \\ &= \|KW\boldsymbol{\beta} - \boldsymbol{\beta}\|^2 + \mathcal{E}[\langle K\boldsymbol{\epsilon}, K\boldsymbol{\epsilon} \rangle].\end{aligned}$$

- Consider unbiased estimation:

- ◇ Observe

$$\mathcal{E}[\hat{\boldsymbol{\beta}}] = \mathcal{E}[KW\boldsymbol{\beta} + K\boldsymbol{\epsilon}] = KW\mathcal{E}[\boldsymbol{\beta}].$$

It is expected that $KW = I_n$.

- ◇ The problem now becomes, given a symmetric and positive definite matrix Q ,

$$\begin{aligned}\text{minimize}_{K \in \mathbb{R}^{n \times m}} \quad & \text{trace } KQK^\top \\ \text{subject to} \quad & KW = I_n.\end{aligned}$$

- ◇ This is in the form of a standard minimum norm problem.

- The problem has a closed form solution.

- ◇ The optimal solution is given by

$$K = (W^\top Q^{-1} W)^{-1} W^\top Q^{-1}.$$

- ◇ The minimum-variance unbiased estimation of $\boldsymbol{\beta}$ is given by

$$\hat{\boldsymbol{\beta}} = (W^\top Q^{-1} W)^{-1} W^\top Q^{-1} \mathbf{y}.$$

- ◇ The special case $Q = I_m$ is the classical least squares problem.

- ▷ The classical least squares solution is providing the unbiased minimum-variance estimate of $\boldsymbol{\beta}$, if the perturbation presented in data is white noise.

- It can be argued that the above solution $\hat{\boldsymbol{\beta}}_i$ is the minimum-variance unbiased estimation of $\boldsymbol{\beta}_i$ for each individual i .

- ◇ This is the true minimum-variance unbiased estimate.

Minimum-Variance Model

- Assume in the linear model

$$\mathbf{y} = W\boldsymbol{\beta} + \boldsymbol{\epsilon}.$$

- ◊ $W \in \mathbb{R}^{m \times n}$ is known.
 - ◊ $\boldsymbol{\epsilon} \in \mathbb{R}^m$ is a random vector with zero mean and covariance $\mathcal{E}(\boldsymbol{\epsilon}\boldsymbol{\epsilon}^\top) = Q$.
 - ◊ $\boldsymbol{\beta}$ is a random vector in \mathbb{R}^n with known statistical information.
 - ◊ \mathbf{y} represents the outcome of inexact measurements in \mathbb{R}^m .
- Want to estimate the unknown random vector $\boldsymbol{\beta} \in \mathbb{R}^n$ based on $\mathbf{y} \in \mathbb{R}^m$ using

$$\hat{\boldsymbol{\beta}} := K\mathbf{y}$$

where K is an unknown matrix in $\mathbb{R}^{n \times m}$.

- The best approximation is measured by minimizing the expected value of the random error, i.e.,

$$\min_{K \in \mathbb{R}^{n \times m}} \mathcal{E}[\|\hat{\boldsymbol{\beta}} - \boldsymbol{\beta}\|^2]$$

Minimum-Variance Estimate

- Assume $(\mathcal{E}[\mathbf{y}\mathbf{y}^T])^{-1}$ exists. Then the minimum-variance estimator of β is given by

$$\hat{\beta} = \mathcal{E}[\beta\mathbf{y}^T](\mathcal{E}[\mathbf{y}\mathbf{y}^T])^{-1}\mathbf{y}.$$

◇ The estimate is independent of W and ϵ .

- Proof Is Interesting!

◇ Write K in rows, i.e., $K = [\mathbf{k}_1^\top, \dots, \mathbf{k}_n^\top]^\top$.

◇ $\mathcal{E}[\|\hat{\beta} - \beta\|^2] = \sum_{i=1}^n \mathcal{E}[(\hat{\beta}_i - \beta_i)^2] = \sum_{i=1}^n \mathcal{E}[(\mathbf{k}_i^\top \mathbf{y} - \beta_i)^2]$.

▷ Suffices to consider each individual term.

◇ Let $f(\mathbf{y}, \beta_i)$ denote the (unknown) joint pdf of \mathbf{y} and β_i .

▷ Define

$$\begin{aligned} g(\mathbf{k}_i) &:= \mathcal{E}[(\mathbf{y}^\top \mathbf{k}_i - \beta_i)^2] \\ &= \int \int (\mathbf{y}^\top \mathbf{k}_i - \beta_i)^2 f(\mathbf{y}, \beta_i) d\mathbf{y} d\beta_i. \end{aligned}$$

▷ Necessary condition is $\nabla g(\mathbf{k}_i) = 0$.

◇ Easy to see

$$\begin{aligned}\frac{\partial g}{\partial \mathbf{k}_{i,j}} &= \int \int 2(\mathbf{y}^\top \mathbf{k}_i - \beta_i) \mathbf{y}_j f(\mathbf{y}, \beta_i) d\mathbf{y} d\beta_i \\ &= 2\mathcal{E}[(\mathbf{y}^\top \mathbf{k}_i - \beta_i) \mathbf{y}_j].\end{aligned}$$

◇ Rewrite the necessary condition as

$$\begin{aligned}\mathcal{E}[\mathbf{y}(\mathbf{y}^\top \mathbf{k}_i - \beta_i)] &= 0, \quad (\text{for each } i) \\ \mathcal{E}[\mathbf{y}\mathbf{y}^\top]K^\top &= \mathcal{E}[\mathbf{y}\beta^\top], \quad (\text{in matrix form}) \\ K &= \mathcal{E}[\beta\mathbf{y}^\top](\mathcal{E}[\mathbf{y}\mathbf{y}^\top])^{-1}.\end{aligned}$$

• The estimate so far is *biased*, unless $\mathcal{E}[\beta] = \mathcal{E}[\mathbf{y}] = 0$.

• In the general case where $\mathcal{E}[\beta] = \beta_0$ and $\mathcal{E}[\mathbf{y}] = \mathbf{y}_0$,

◇ The estimate should assume the form

$$\hat{\beta} := K\mathbf{y} + \mathbf{b}.$$

◇ The minimum-variance estimate is given by

$$\hat{\beta} = \beta_0 + \mathcal{E}[(\beta - \beta_0)(\mathbf{y} - \mathbf{y}_0)^\top](\mathcal{E}[(\mathbf{y} - \mathbf{y}_0)(\mathbf{y} - \mathbf{y}_0)^\top])^{-1}(\mathbf{y} - \mathbf{y}_0).$$

Equivalent Formulas

- Assume that

$$\begin{aligned}\mathcal{E}[\boldsymbol{\epsilon}\boldsymbol{\epsilon}^\top] &= Q \in \mathbb{R}^{m \times m}, \\ \mathcal{E}[\boldsymbol{\beta}\boldsymbol{\beta}^\top] &= R \in \mathbb{R}^{n \times n}, \\ \mathcal{E}[\boldsymbol{\epsilon}\boldsymbol{\beta}^\top] &= 0 \in \mathbb{R}^{m \times n}.\end{aligned}$$

- Then the minimum-variance estimate can be written as

$$\begin{aligned}\hat{\boldsymbol{\beta}} &= \underbrace{RW^\top(WRW^\top + Q)^{-1}}_K \mathbf{y} \\ &= \underbrace{(W^\top Q^{-1}W + R^{-1})^{-1}W^\top Q^{-1}}_K \mathbf{y}.\end{aligned}$$

- ◇ The equivalence can be proved by direct substitution.
- ◇ [Check out the dimension of the inverse matrices involved in the two expressions.](#)

Comparison with Gauss-Markov Estimate

- The Gauss-Markov estimate is

$$\hat{\boldsymbol{\beta}} = (W^\top Q^{-1}W)^{-1} W^\top Q^{-1}\mathbf{y}.$$

- The more subtle minimum-variance estimate is

$$\hat{\boldsymbol{\beta}} = (W^\top Q^{-1}W + R^{-1})^{-1} W^\top Q^{-1}\mathbf{y}.$$

- If $R^{-1} = 0$, the two estimates are identical.
 - ◇ What is meant by $R^{-1} = 0$?
 - ◇ *Infinite* variance of $\boldsymbol{\beta}$ in the more subtle estimate means that we have *absolutely no a priori knowledge* of $\boldsymbol{\beta}$ at all.
- When $\boldsymbol{\beta}$ is considered as a random variable, the size m of observations \mathbf{y} does not need to be large.
 - ◇ $(WRW^\top + Q)^{-1}$ exists so long as Q is positive positive definite.
 - ◇ Every new measurement simply provides additional information which may modify the original estimate.

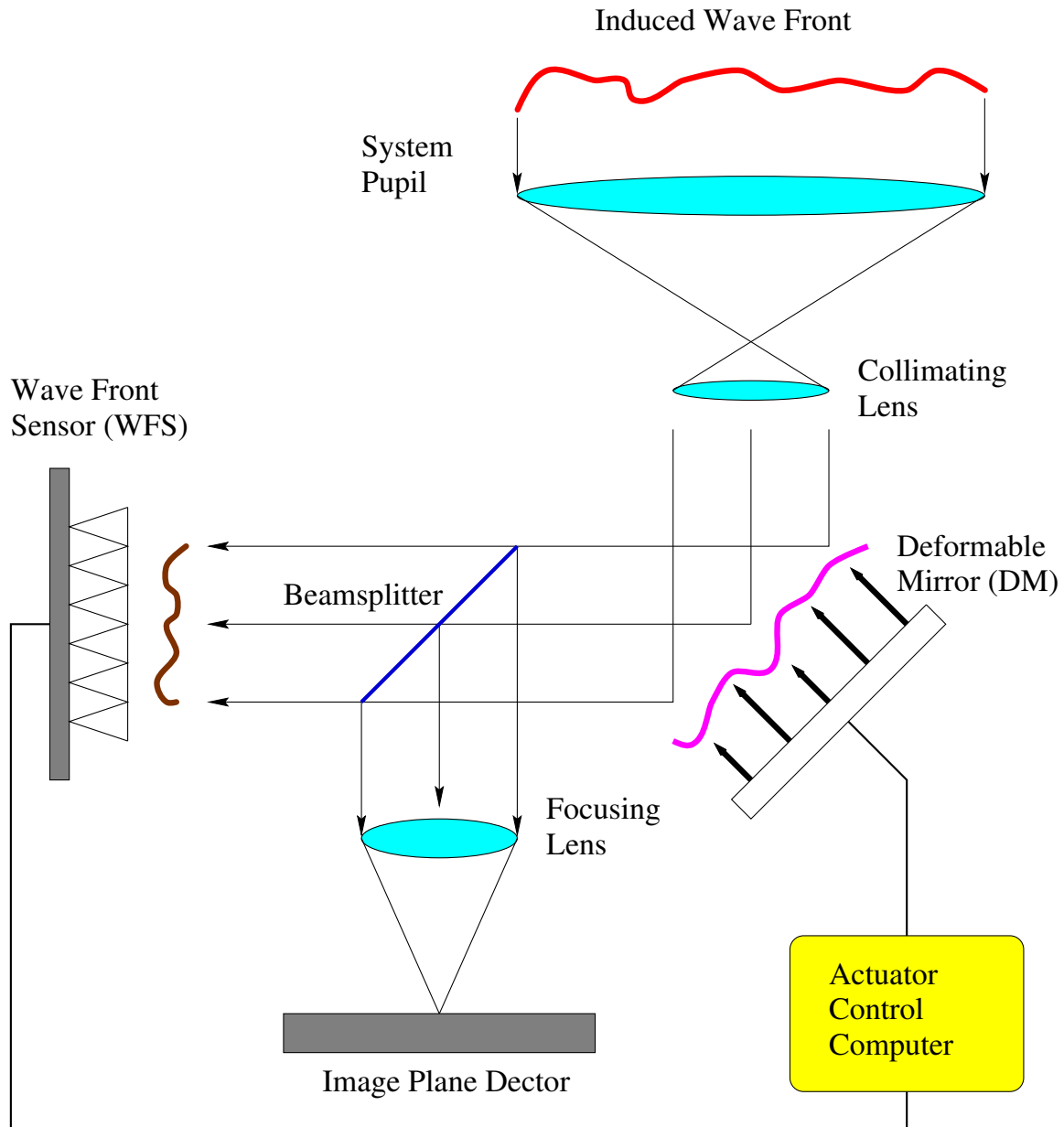
Application to Adaptive Optics

- Imaging through the Atmosphere
- Adaptive Optics System
 - ◇ Basic Relationships
 - ◇ Open-loop Model
 - ◇ Closed-loop Model
- Adaptive Optics Control
 - ◇ An Ideal Control
 - ◇ An Inverse Problem
 - ◇ Temporary Latency
- Numerical Illustration

Atmospheric Imaging Computation

- Purpose:
 - ◇ To compensate for the degradation of astronomical image quality caused by the effects of atmospheric turbulence.
- Two stages of approach:
 - ◇ Partially nullify optical distortions by a deformable mirror (DM) operated from a closed-loop adaptive optics (AO) system.
 - ◇ Minimize noise or blur via off-line post-processing deconvolution techniques (not this talk).
- Challenges:
 - ◇ Atmospheric turbulence can only be measured adaptively.
 - ◇ Need theory to pass atmospheric measurements to knowledge of actuating the DM.
 - ◇ Require fast performance of large-scale data processing and computations.

A Simplified AO System



Basic Notation

- Three quantities:
 - ◇ $\phi(t)$ = turbulence-induced phase profile at time t .
 - ◇ $a(t)$ = deformable mirror (DM) actuator command at time t .
 - ◇ $s(t)$ = wavefront slope sensor (WFS) measurement at time t and with no correction.
- Two transformations:
 - ◇ H := transformation from actuator commands to resulting phase profile adjustments.
 - ◇ G := transformation from actuator commands to slope sensor measurement adjustments.

From Actuator to DM Surface

- H is used to describe the DM surface change due to the application of actuators.
- $r_i(\vec{x})$ = influence function on the DM surface at position \vec{x} with an unit adjustment to the i th actuator.
- Assuming m actuators and linear response of actuators to the command, model the DM surface by

$$\hat{\phi}(\vec{x}, t) = \sum_{i=1}^m a_i(t) r_i(\vec{x}).$$

◇ Sampled at n DM surface positions, can write

$$\boxed{\hat{\phi}(t) = Ha(t)}$$

- ▷ $H = (r_i(\vec{x}_j)) \in R^{n \times m}$.
- ▷ $\hat{\phi}(t) = [\hat{\phi}(\vec{x}_1, t), \dots, \hat{\phi}(\vec{x}_n, t)]^T \in R^n$ = discrete corrected phase profile at time t .

From Actuator to WFS Measurement

- G is used to describe the WFS slope measurement associated with the actuator command a .
- Consider the H-WFS model where

$$s_j(t) := - \int d\vec{x} (\nabla W_j(\vec{x}) \phi(\vec{x}, t)), \quad j = 1, \dots, \ell.$$

◊ W_j = given specifications of j th subaperture.

- The measurement corresponding to $\hat{\phi}(\vec{x}, t)$ would be

$$\hat{s}_j(t) = \sum_{i=1}^m \underbrace{\left(- \int d\vec{x} (\nabla W_j(\vec{x}) r_i(\vec{x})) \right)}_{G_{ji}} a_i(t).$$

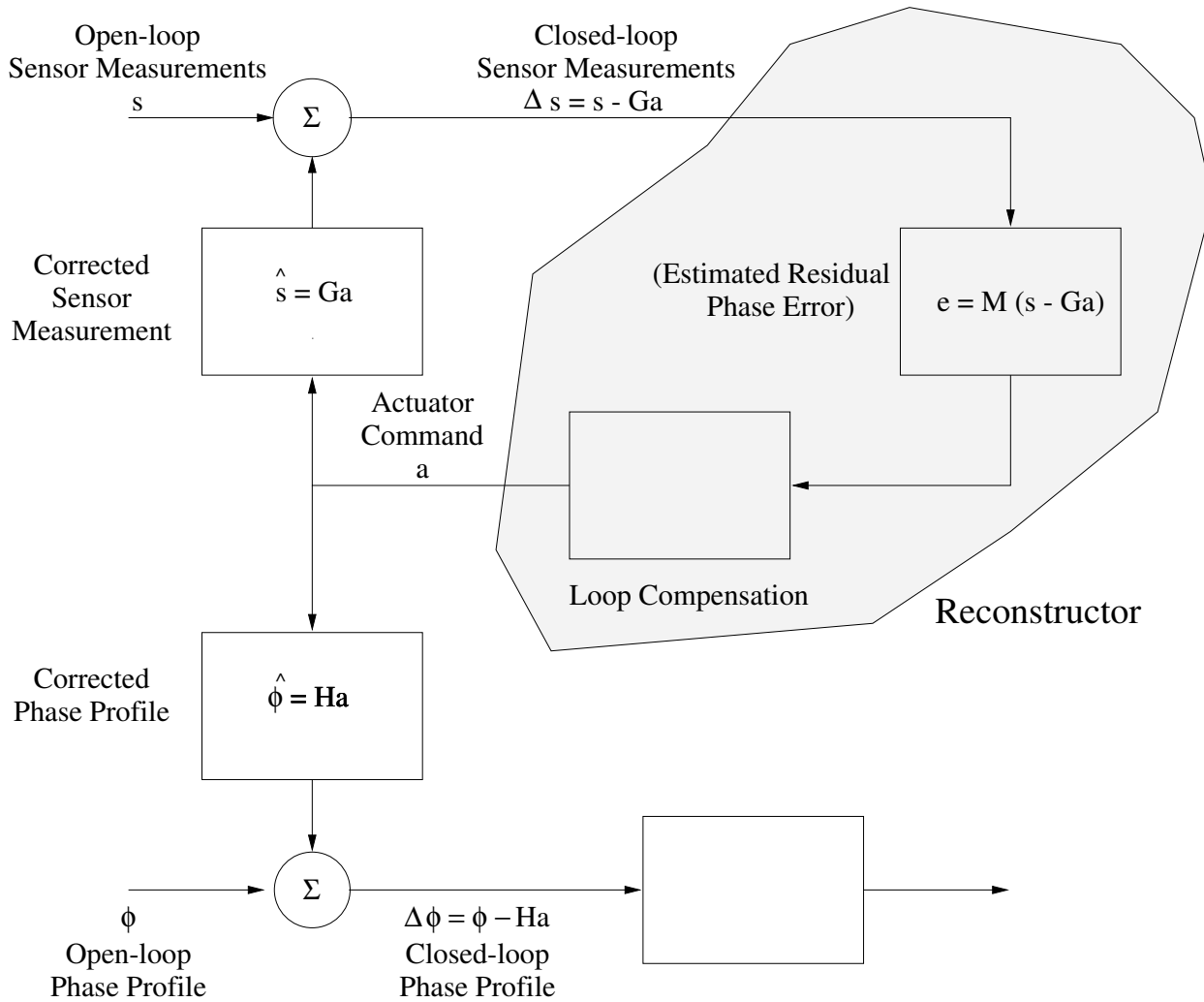
◊ Can write

$$\boxed{\hat{s}(t) = Ga(t)}$$

where $G = [G_{ij}] \in R^{\ell \times m}$.

- ◊ The DM actuators are *not* capable of producing the exact wavefront phase $\phi(\vec{x}, t)$ due to its finiteness of degrees of freedom. So $\hat{s} = Ga$ is not an exact measurement.

A Closed-loop AO Control Model



What is Available?

- Two residuals that are available in a *closed-loop* AO system:
 - ◇ $\Delta\phi(t) := \phi(t) - Ha(t)$
 - ▷ Represents the residual phase error remaining after the AO correction.
 - ▷ Also means instantaneous closed-loop wavefront distortion at time t .
 - ◇ $\Delta s(t) := s(t) - Ga(t)$
 - ▷ Represents feedback applied to $s(t)$ by DM actuator adjustment.
 - ▷ Also means *observable* wavefront sensor measurement at time t .
- In practice, there is a servo lag or delay in time Δt , i.e., it is likely
 - ◇ $\Delta\phi(t) := \phi(t) - Ha(t - \Delta t)$.
 - ◇ $\Delta s(t) := s(t) - Ga(t - \Delta t)$.

Thus the data collected are not perfect.

Open-loop Model

- Assume a linear relationship between open-loop WFS measurement s and turbulence-induced phase profile ϕ :

$$\boxed{s = W\phi + \epsilon}. \quad (1)$$

- ◇ ϵ = measurement noise with mean zero.
 - ◇ In the H-WFS model, W represents a quadrature of the integral operator evaluated at designated positions \vec{x}_j , $j = 1, \dots, n$.
- Want to estimate ϕ using $\tilde{\phi}$ from the model

$$\tilde{\phi} = E_{open}s$$

so that the variance

$$\mathcal{E}[\|\phi - \tilde{\phi}\|^2]$$

is minimized.

- ◇ The wave front reconstruction matrix E_{open} is given by

$$E_{open} = \mathcal{E}[\phi s^T](\mathcal{E}[s s^T])^{-1}.$$

- ◇ For unbiased estimation, need to enforce the condition that $E_{open}W = I$.

Closed-loop Model

- For the H-WFS model, it is reasonable to assume the relationship

$$WH = G. \quad (2)$$

- Then

$$\begin{aligned} s &= W\phi + \epsilon \\ &= W(Ha + \Delta\phi) + \epsilon \\ &= WHa + (W\Delta\phi + \epsilon). \end{aligned}$$

It follows that

$$\boxed{\Delta s = W\Delta\phi + \epsilon}. \quad (3)$$

- ◇ The closed-loop relationship (3) is identical to the open-loop relationship (1).
- Can estimate the residual phase error $\Delta\phi(t)$ using $\Delta\tilde{\phi}(t)$ from the model

$$\Delta\tilde{\phi} = E_{closed}\Delta s$$

- ◇ E_{closed} = wavefront reconstruction matrix.
 - ◇ For unbiased estimation, it requires that $E_{closed}W = I$. Hence

$$E_{closed}G = e_{closed}(WH) = H.$$

Actuator Control

- An Ideal Control:

- ◇ $\Delta\phi$ = residual error after DM correction by current command a_c .

- ◇ New command a_+ should reduce the residual error, i.e., want to

$$\min_a \|Ha - \phi\|.$$

- ◇ Define $\Delta a := a_+ - a_c$, then want to

$$\min_{\Delta a} \|H\Delta a - \Delta\phi\|.$$

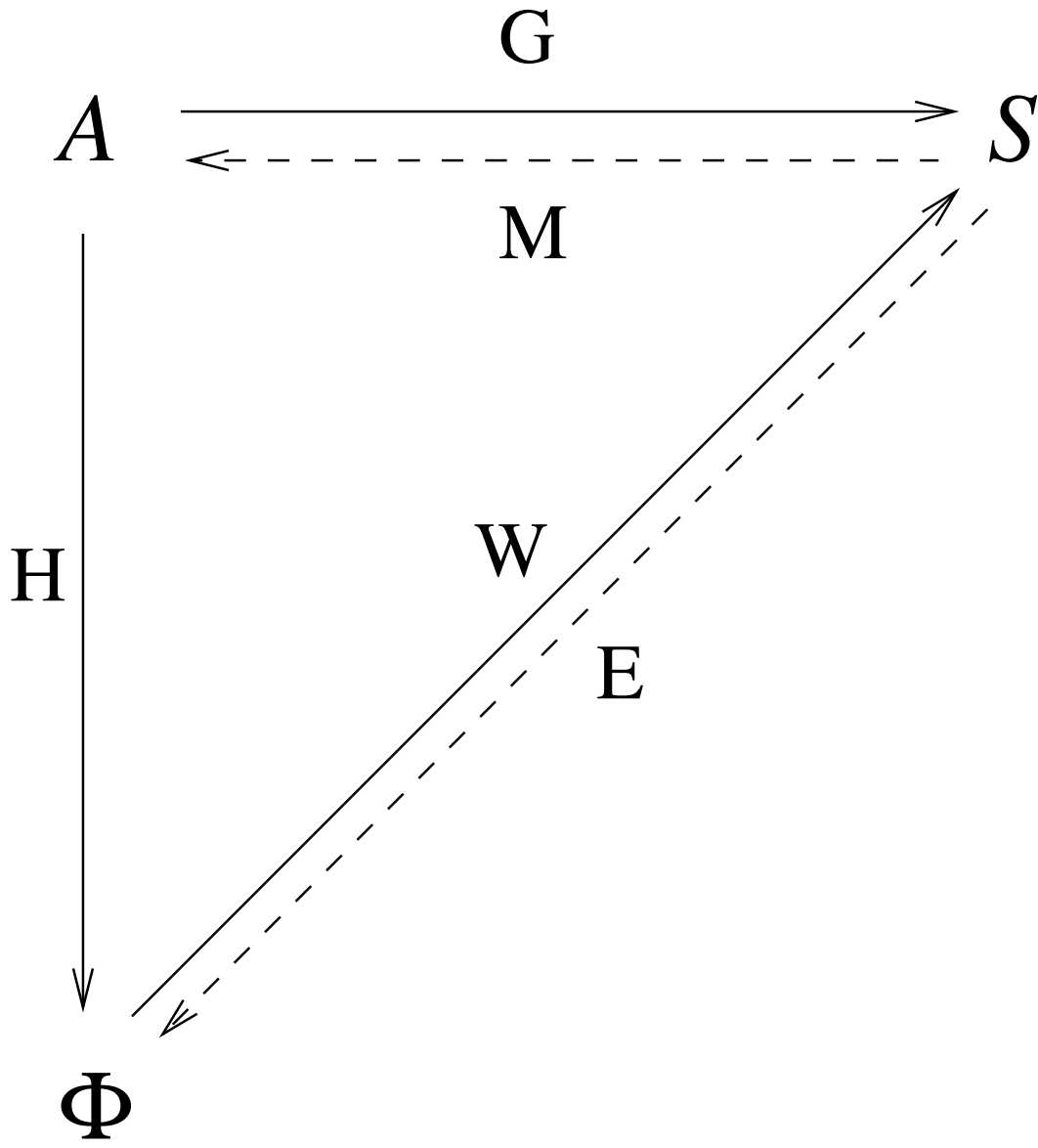
- ◇ But $\Delta\phi$ is not observable directly. It has to be estimated from Δs .

- Estimating Δa directly from Δs :

$$\boxed{\Delta a = M\Delta s}$$

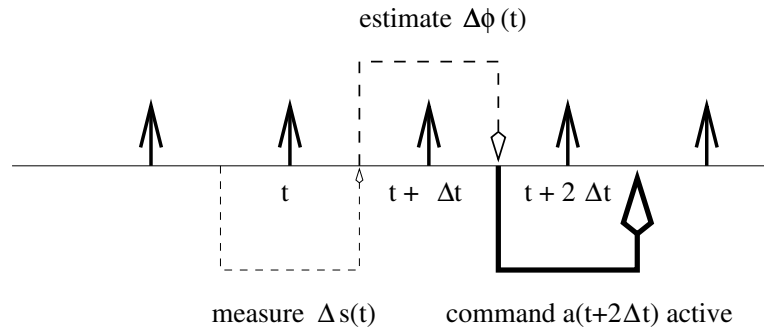
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An Inverse Problem



Actuator Control with Temporary Latency

- Due to finite bandwidth of the control loop, Δs is not immediately available.
- Time line for the scenario of a 2-cycle delay,



- ARMA control scheme:

$$a(t + 2\Delta t) := \sum_{k=0}^p c_k a(t + (1 - k)\Delta t) + \sum_{j=0}^q b_j M_j \Delta s(t - j\Delta t).$$

$$a^{(r+2)} = \sum_{k=0}^p c_k a^{(r+1-k)} + \sum_{j=0}^q b_j M_j \Delta s^{(r-j)}, r = 0, 1, \dots$$

Expected Effect on the AO System

- Suppose

- ◇ $\exp[s(t)]$ is independent of time t throughout the cycle of computation.
- ◇ Matrix $\sum_{j=0}^q b_j M_j$ is of full column rank.

- Then

- ◇ The WFS feedback measurement $\Delta s^{(n)}$ is eventually nullified by the actuators, i.e.,

$$\exp[s] = G \lim_{n \rightarrow \infty} \exp[a^{(n)}].$$

- ◇ The expected residual phase error is inversely related to the expected WFS measurement noise ϵ via

$$0 = W \lim_{n \rightarrow \infty} \exp[\Delta \phi^{(n)}] + \exp[\epsilon].$$

- Compare with the ideal control:

- ◇ Even if $\exp[\epsilon] = 0$, not necessarily $\exp[\|\lim_{n \rightarrow \infty} \Delta \phi_n\|^2]$ will be small because W has non-trivial null space.

Almost Sure Convergence

- Each control $a^{(r+j)}$ is a random variable \implies The control scheme is a stochastic process.
- Each control $A^{(r+j)}$ is also a realization of the corresponding random variable \implies The control scheme is a deterministic iteration.
- Convergence of deterministic iteration on independent random samples \implies Almost sure convergence of stochastic process.
- Need fast convergence:
 - ◇ Stationary statistic is not realistic.
 - ◇ Atmospheric turbulence changes rapidly.
 - ◇ Can only assume stationary statistic for a short period of time.

- Define

$$\mathbf{a}_{r+2} := [a^{(r+2)}, a^{(r+1)}, \dots, a^{(r-q+1)}]^T, \quad r = 0, 1, \dots$$

$$\mathbf{b} := \left[\sum_{j=0}^q b_j M_j G s', 0, \dots, 0 \right]^T.$$

- The ARMA scheme becomes

$$\mathbf{a}_{r+2} = A\mathbf{a}_{r+1} + \mathbf{b}$$

where A is the $m(q+2) \times m(q+2)$ matrix

$$A := \begin{bmatrix} c_0 I_m & c_1 I_m - b_0 M_1 G & \dots & c_{q+1} I_m - b_q M_q G \\ I_m & 0 & & 0 \\ 0 & I_m & & \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & I_m & 0 \end{bmatrix}$$

- Almost convergence \iff Spectral radius $\rho(A)$ of A is less than one.
- Asymptotic convergence factor is precisely $\rho(A)$.

Numerical Simulation

- Consider the 2-cycle delay scheme

$$a(t + 2\Delta t) = a(t + \Delta t) + 0.6H^\dagger W^\dagger \Delta s(t).$$

- Test data:

surface positions n	=	5
number of actuators m	=	4
number of subapertures ℓ	=	3
size of random samples z	=	2500
H	=	$rand(n, m)$
W	=	$rand(\ell, n)$
G	=	WH
L_ϕ	=	$rand(n, n)$
L_ϵ	=	$diag(rand(\ell, 1))$
μ_ϕ	=	$zeros(n, 1)$
μ_ϵ	=	$zeros(\ell, 1)$

- Random samples:

$$\begin{aligned}\phi &= \mu_\phi * ones(1, z) + L_\phi * randn(n, z), \\ \epsilon &= \mu_\epsilon * ones(1, z) + L_\epsilon * randn(\ell, z).\end{aligned}$$